

Quantum Limited Detectors for Astronomical Receivers

Ghassan Yassin

Astrophysics, University of Oxford, UK

Dublin, 2019

Quantum (Heterodyne) Detectors



- SIS Mixers at Millimere Wave lengths
 - On Chip Arrays
 - Harmonics Multipliers
- THz Mixers
 - ALMA Band 10, band 11 SIS mixers
 - Understanding the physics of SIS mixers above the gap
- Superconducting Parametric Amplifiers
 - Investigation of operating temperature and losses
 - IF amplifiers for SIS mixers
 - Wide band millimeter amplifiers.

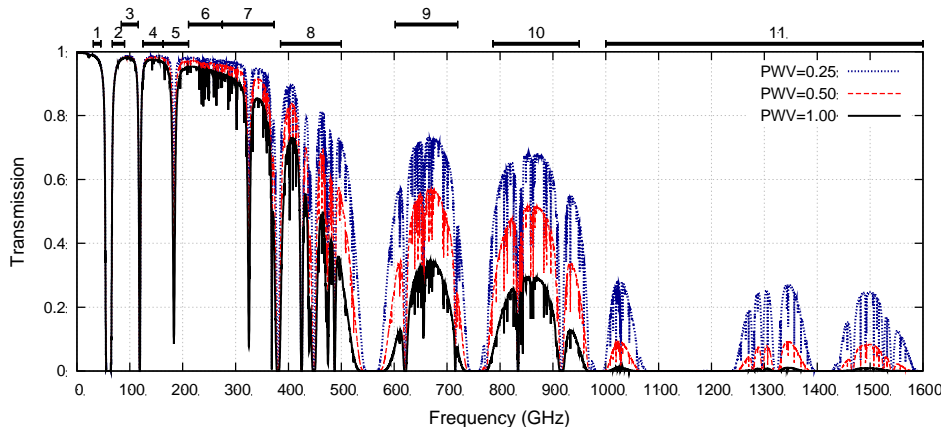
Oxford Quantum Detectors



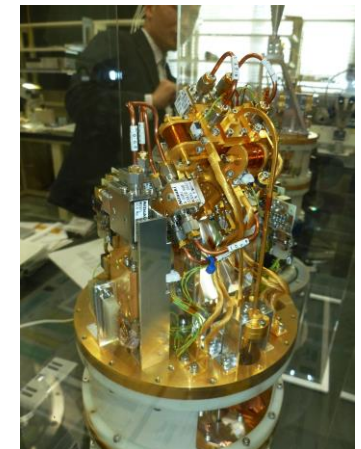
- Group
 - Ghassan, Boon Kok Tan, John Garret, Rik Elliott, Alessandro Traini, Kitti Ratter.
 - Two PhD students.
- Funding
 - STFC Consortium grant (with Cambridge and RAL)
 - Horizon 2020 (RadioNet)
 - ERC Grant+ MERAC fund (SPA)
- Collaboration
 - RAL (LO sources, Mechanical Engineering, Measurements,...)
 - Cambridge (superconducting devices)
 - Paris Observatory (SIS devices Faouzi Boussaha)
 - Harvard-Smithsonian CfA (dual polarization receiver for SMA).



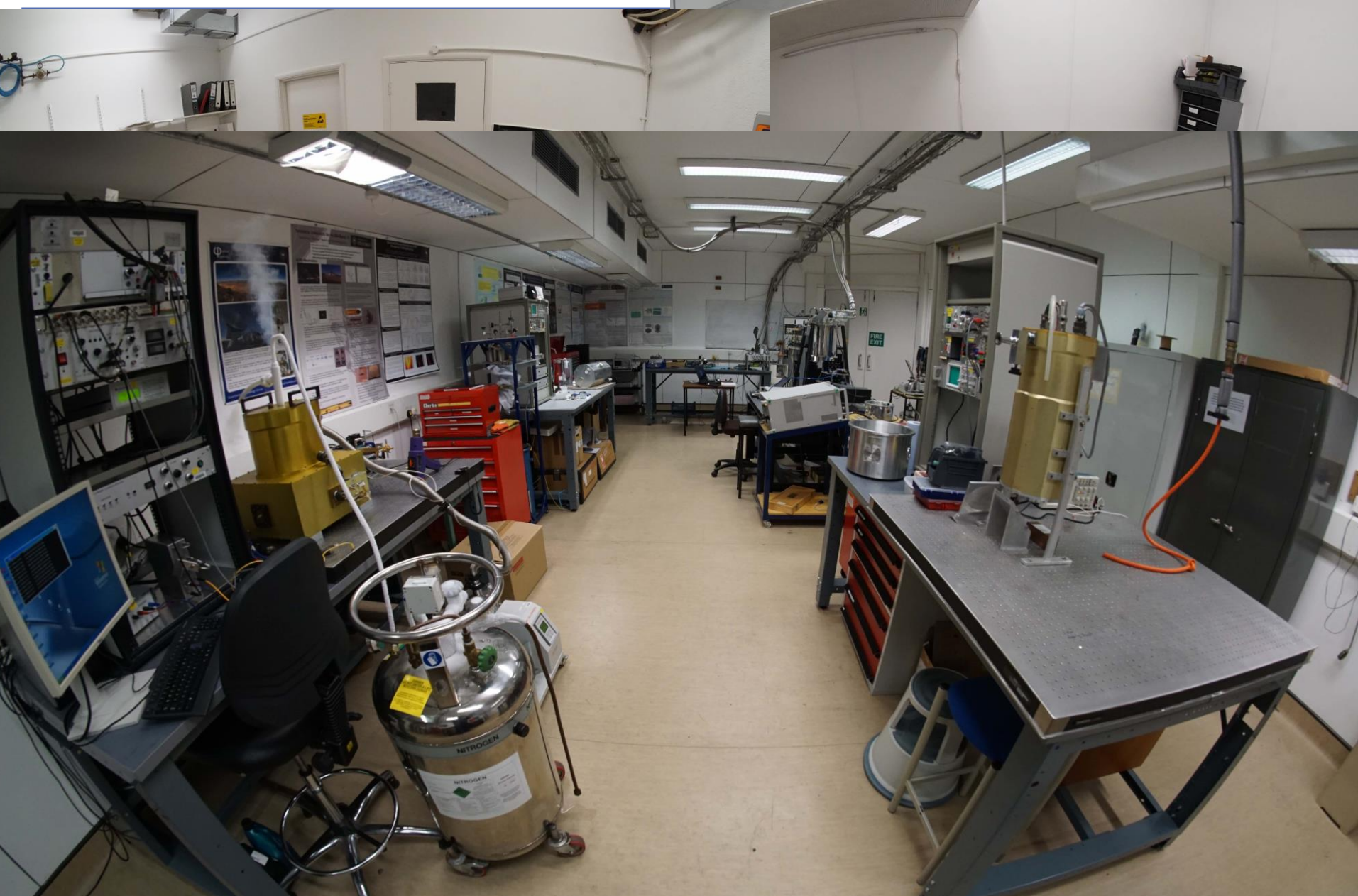
ALMA Receiver Cabin



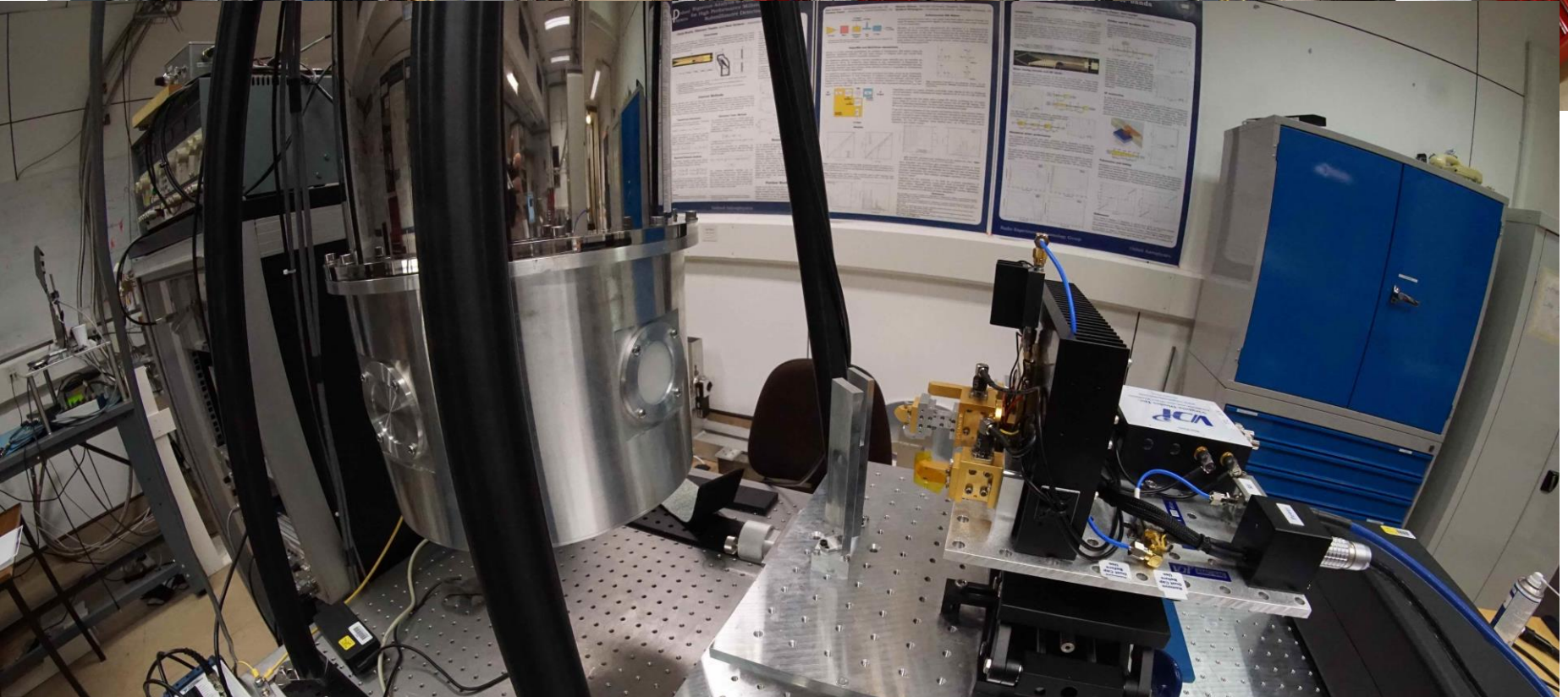
Band	Frequency range (GHz)	Target Receiver Noise Temperature (K)		Detector technology
		80% RF band	At any RF frequency	
1	31-45	17	26	HEMT ^a
2	67-90	30	47	HEMT
3*	84-116	37	60	SIS
4	125-163	51	82	SIS
5	162-211	65	105	SIS
6*	211-275	83	136	SIS
7*	275-373	147	219	SIS
8	385-500	196	292	SIS
9*	602-720	175	261	SIS
10	787-950	230	344	SIS



Lab Photo

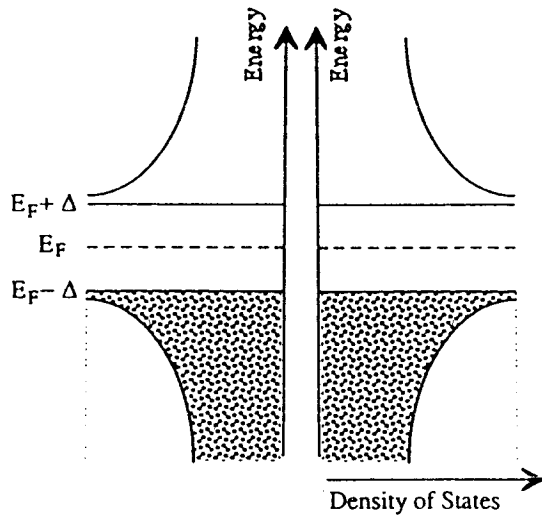


Lab Photo

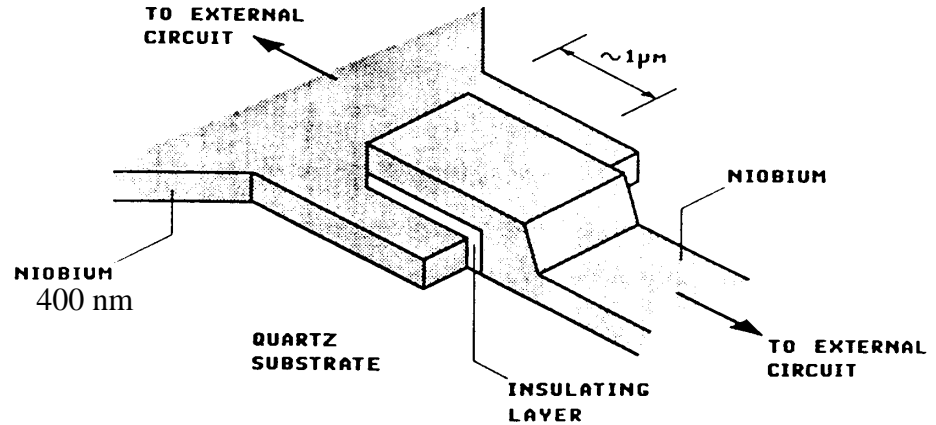


The SIS Tunnel Junction

The SIS (superconductor-insulator-superconductor) is $\sim 1\mu\text{m}^2$ sandwich of two superconductors separated by a thin (20 \AA) insulator



The semiconductor picture



The SIS Nb tunnel junction

quasiparticles are normal electrons with energy and density of states:

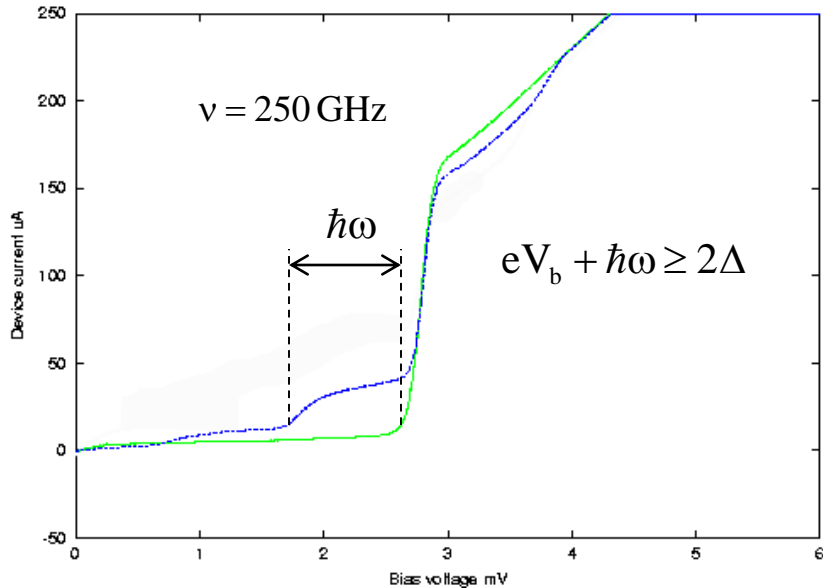
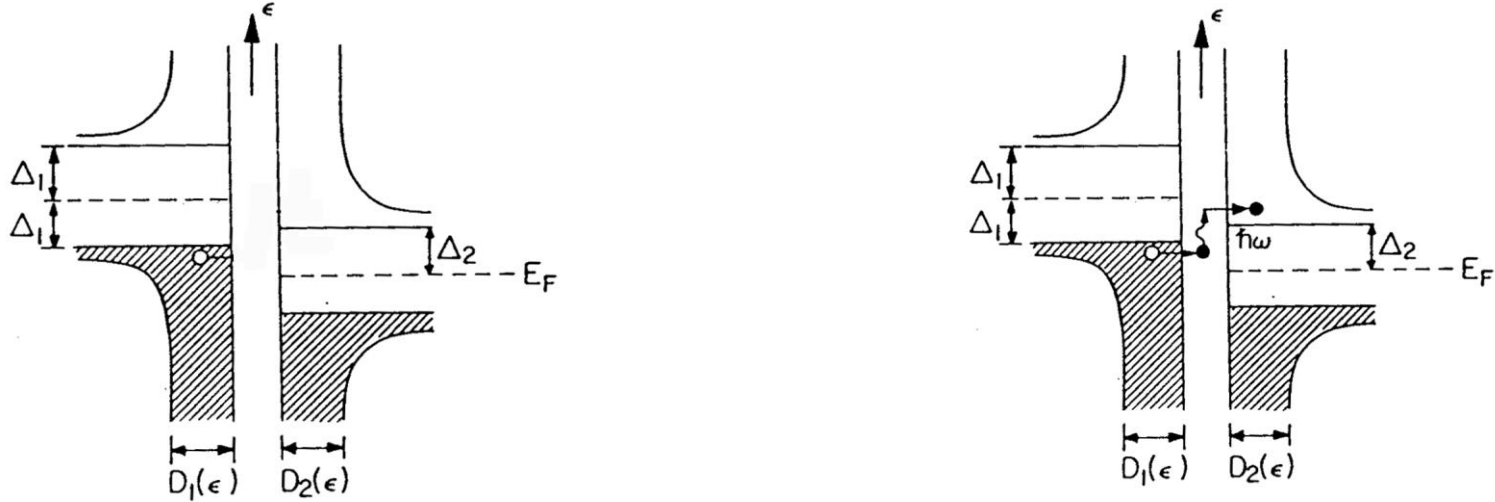
$$E_k = (\varepsilon_k^2 + \Delta^2)^{\frac{1}{2}}$$

$$\rho(E) = \begin{cases} N(0) \frac{|E|}{(E^2 - \Delta^2)^{\frac{1}{2}}} & \text{for } |E| > \Delta \\ 0 & \text{Otherwise} \end{cases}$$

SIS Tunnel Junctions



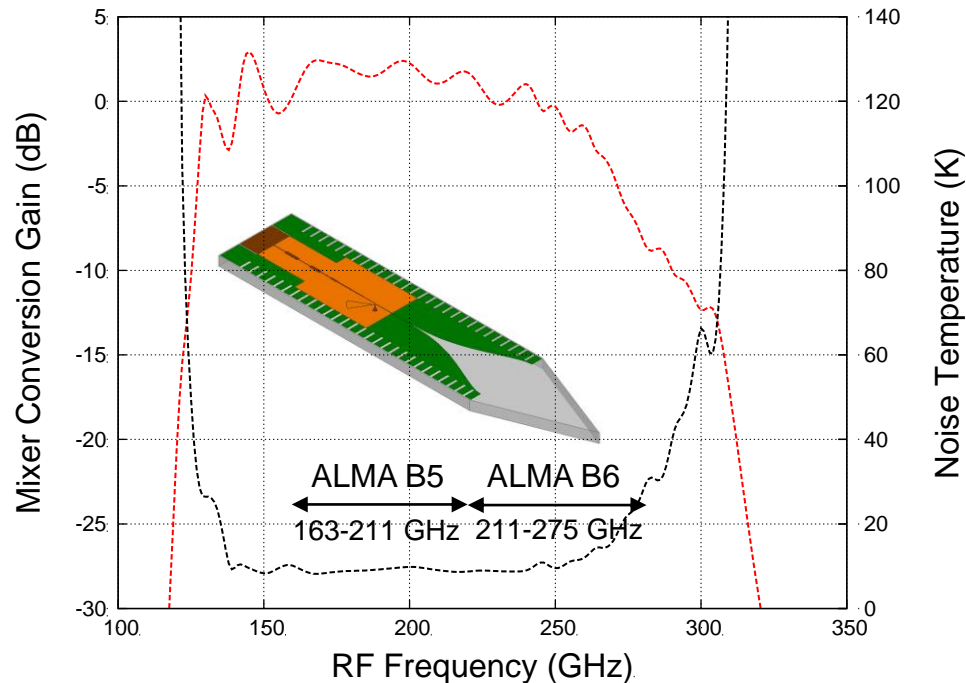
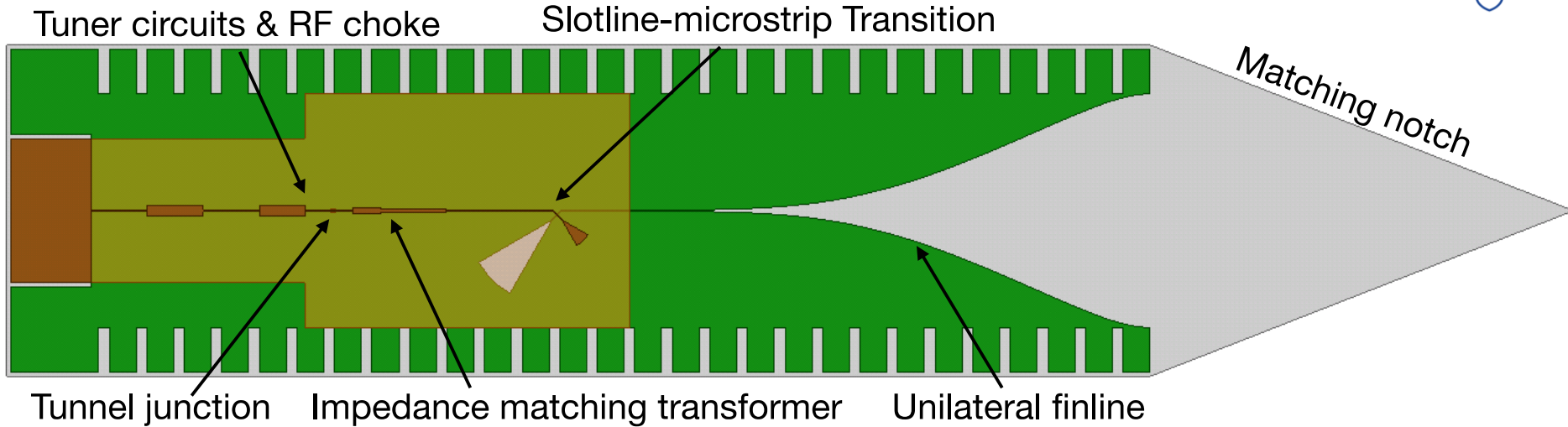
Credit: Dr. Faot



$$E_k = (\epsilon_k^2 + \Delta^2)^{\frac{1}{2}}$$

$$\rho(E) = \begin{cases} N(0) \frac{|E|}{(E^2 - \Delta^2)^{\frac{1}{2}}} & \text{for } |E| > \Delta \\ 0 & \text{Otherwise} \end{cases}$$

Full RF Mixer Chip

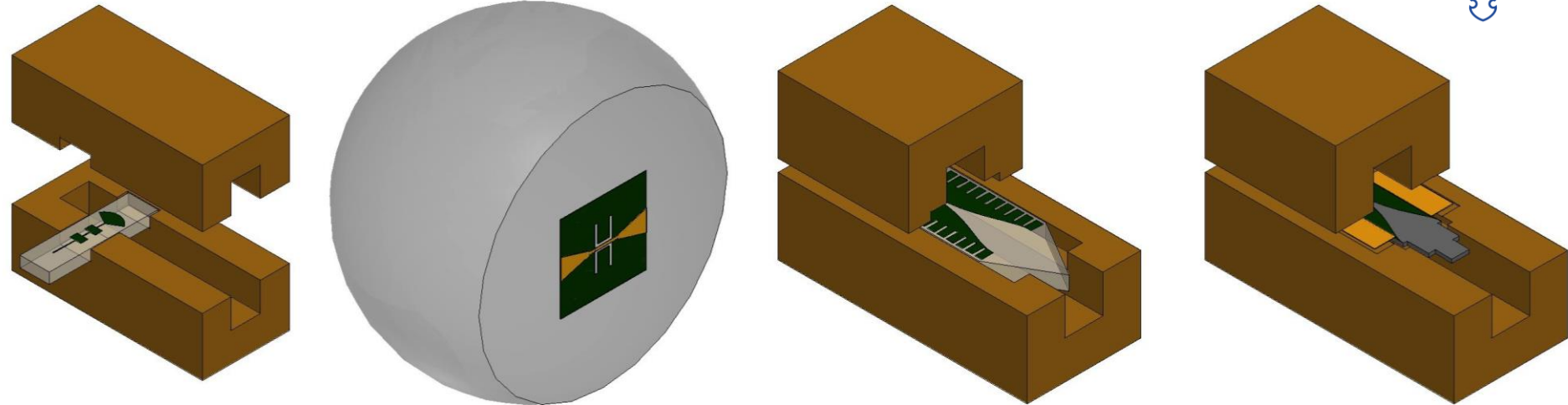


Mixer designs based on:

- 15 ohm, 120 fF, 1.5 μm^2 junction
- 350 nm Nb / 490 nm SiO / 200 nm Nb
- 100 μm Quartz substrate
- Smallest feature 2.5 μm

Optimised using HFSS to include **3D electromagnetic effects** and **superconductivity**

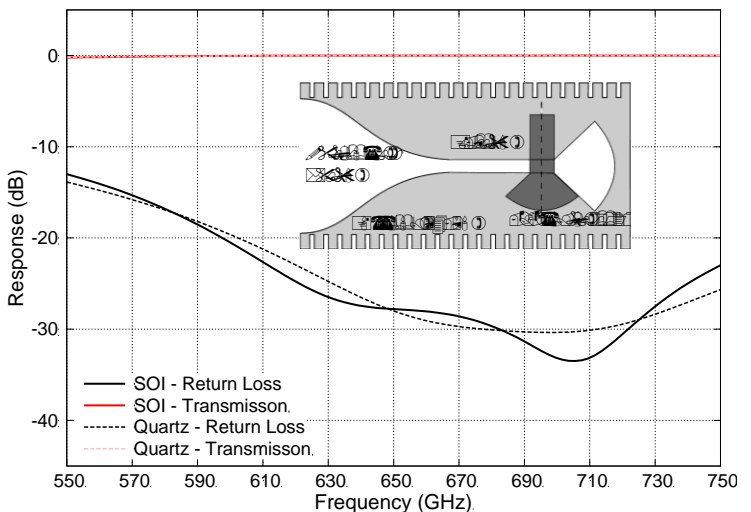
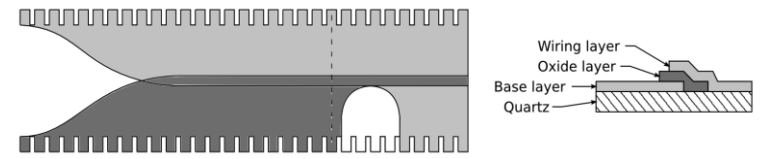
Waveguide to Planar Circuit Transitions



1. Radial Probe

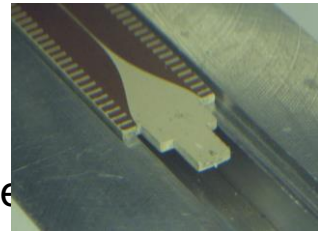
2. Twin-slot Antenna

3. Unilateral Finline Taper



Unilateral Finline Taper

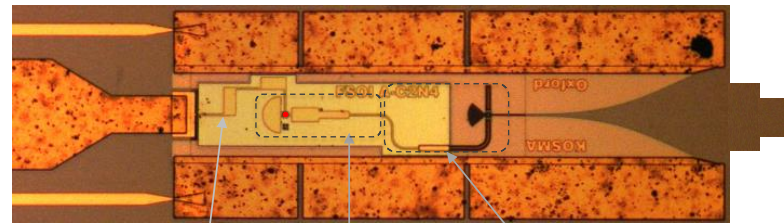
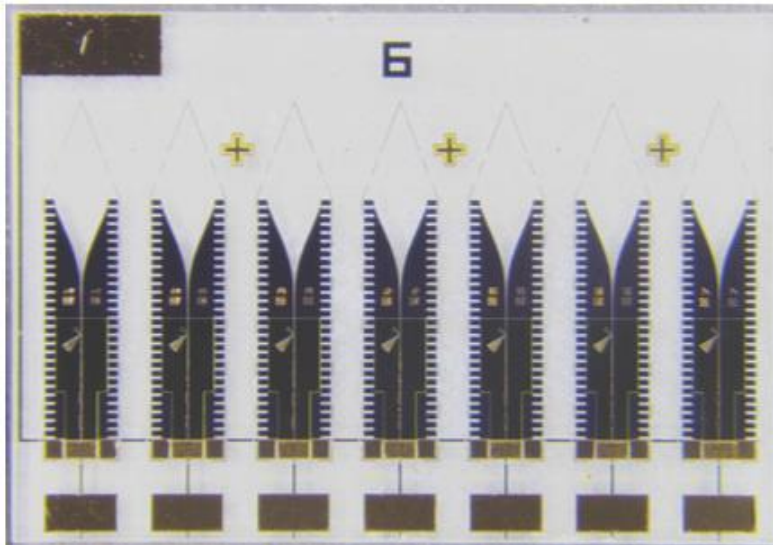
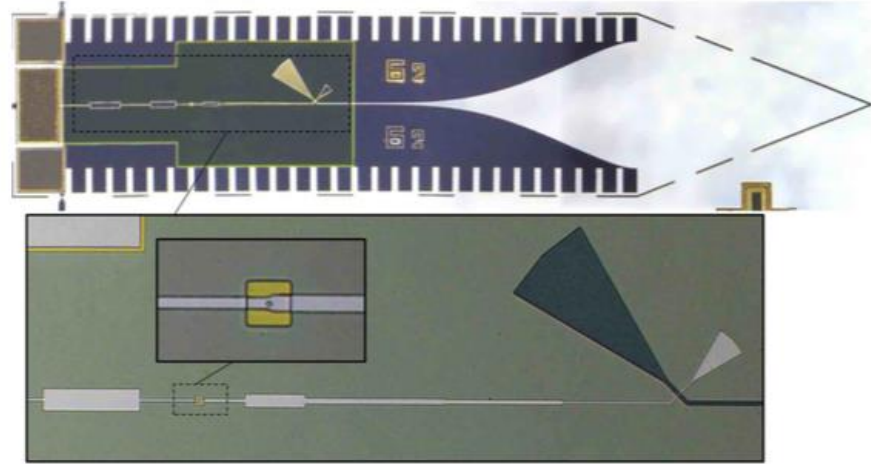
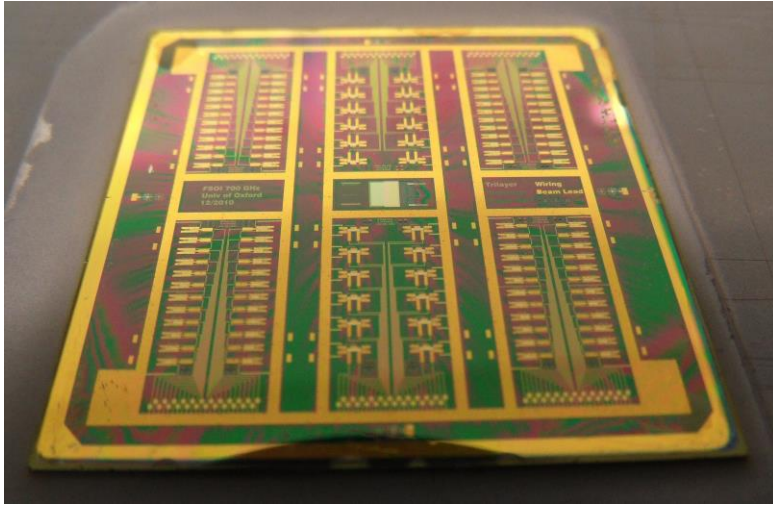
- Oriented along waveguide axis
- No backshort, minimal alignment issues
- Natural bandpass filter, prevent IF signal leakage
- Simple mixer block
- Wide bandwidth, smooth tapering of waveguide impedance to slotline
- Large substrate area for circuit integration
- Fully on-chip fabrication, high frequency



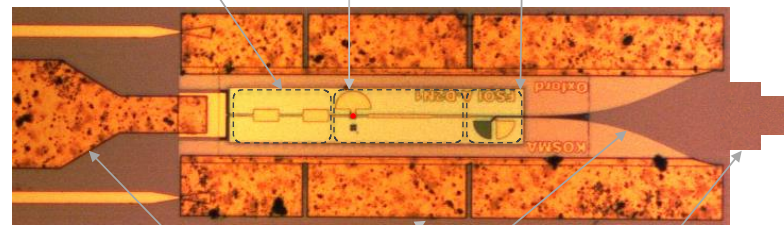
Band 5/6 & Band 9 Mixers



Credit: Dr. John Garrett (Oxford University) & Cologne University



CPW-mixer
RF choke
Tuner circuit
Finline-microstrip transition

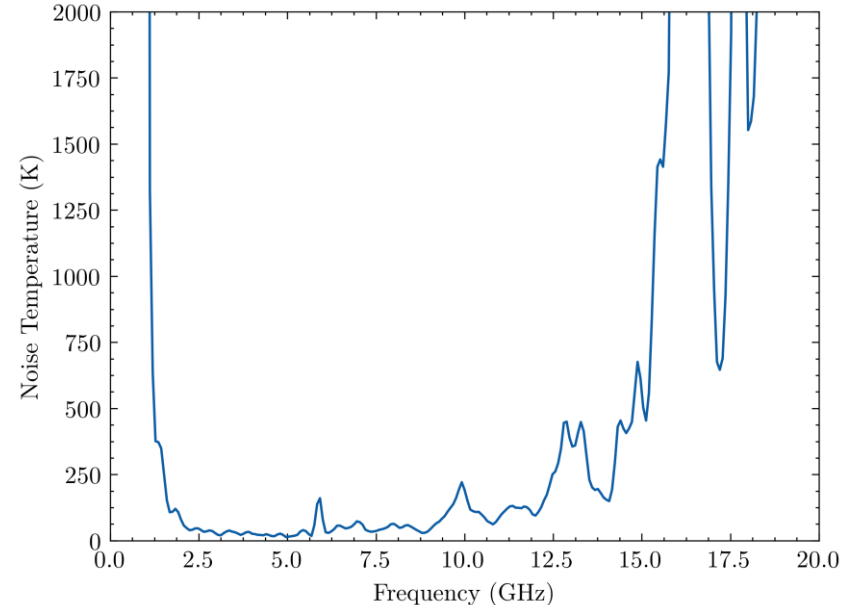
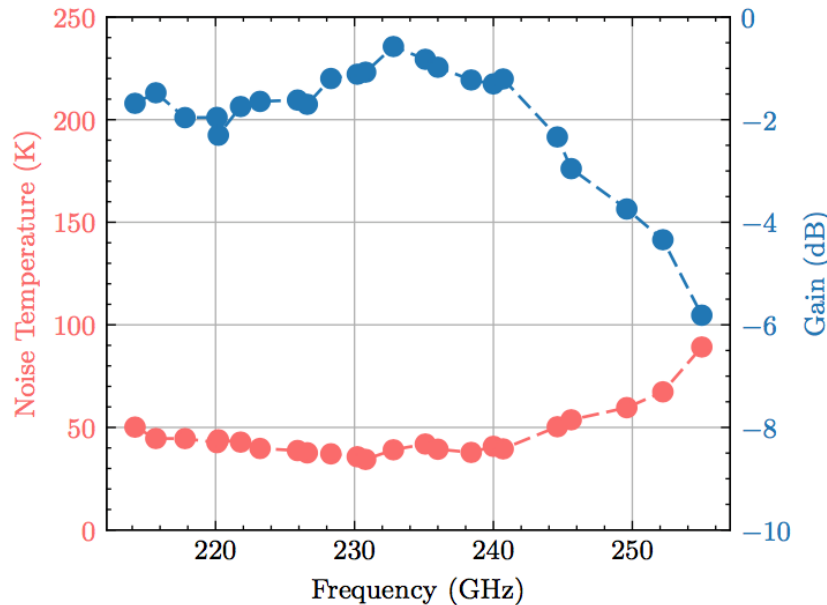
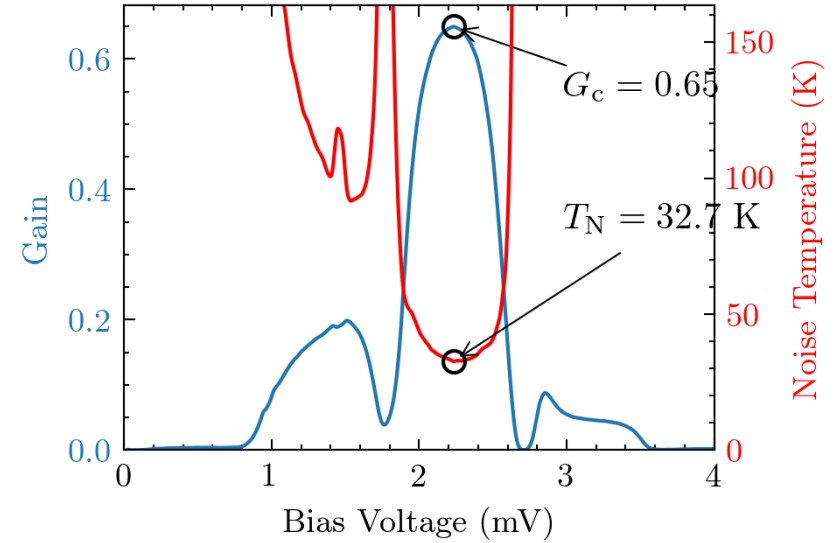
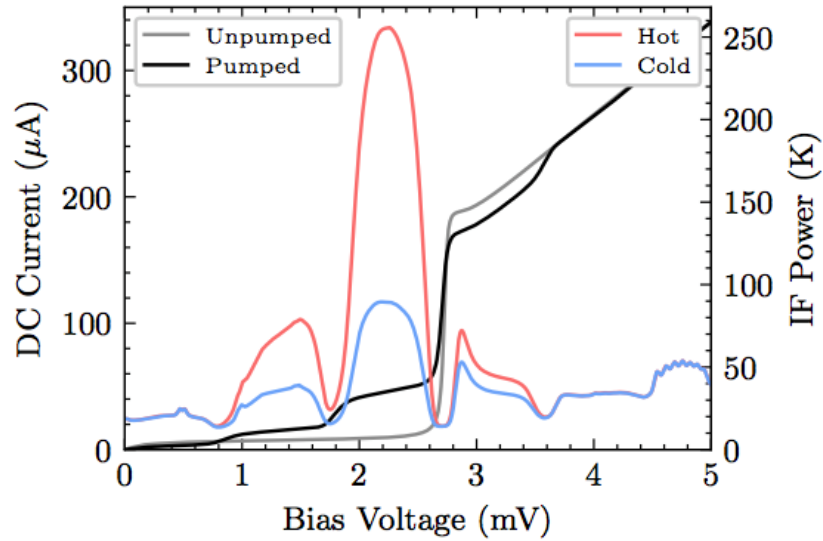


Direct-mixer
Gold beam leads (IF signal)
Gold beam leads (Ground)
Unilateral finline taper
Multi-stage matching notch

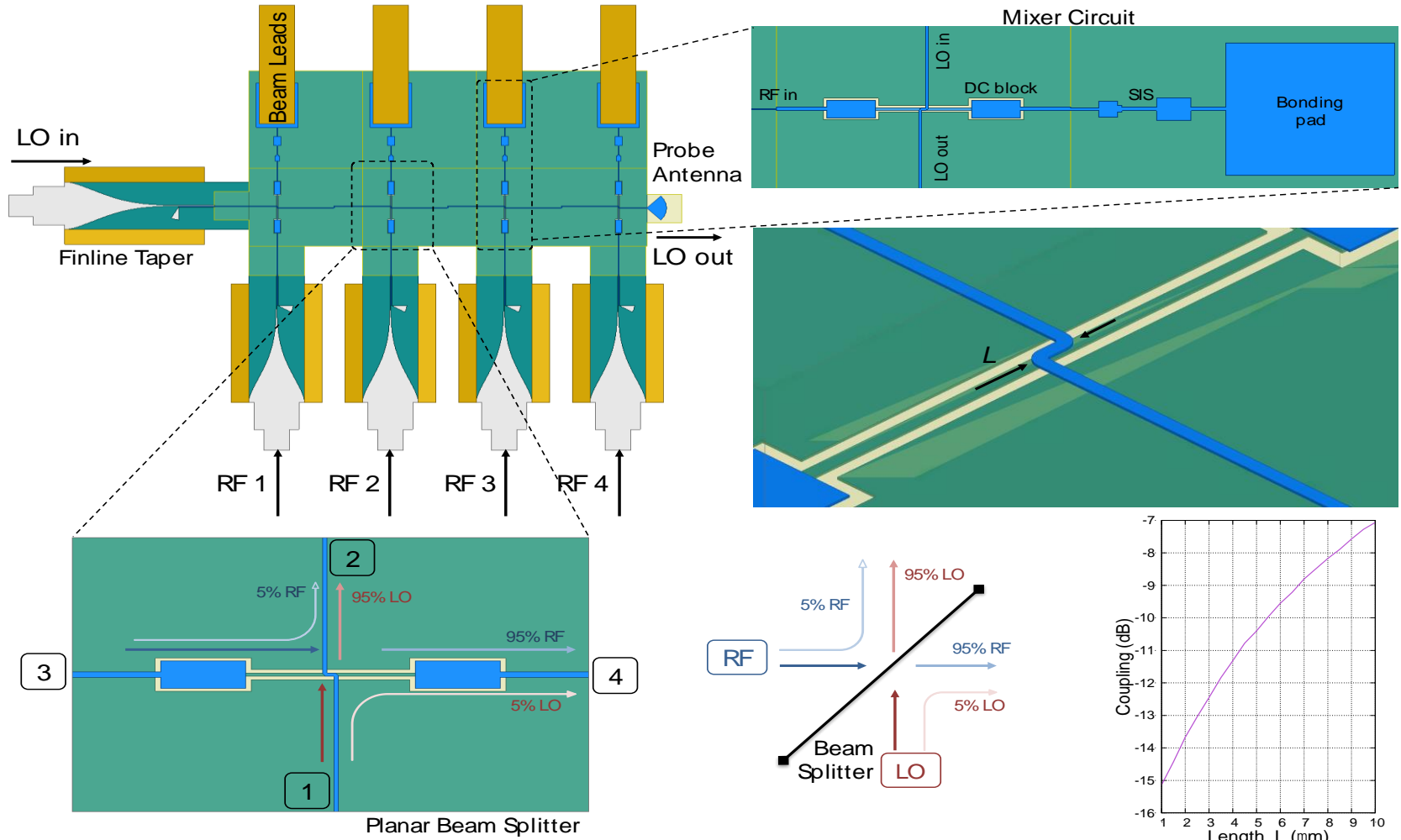
Measured Characteristics



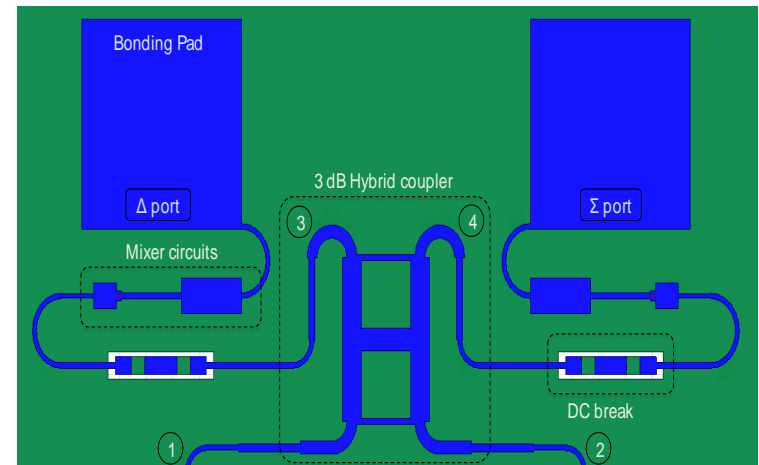
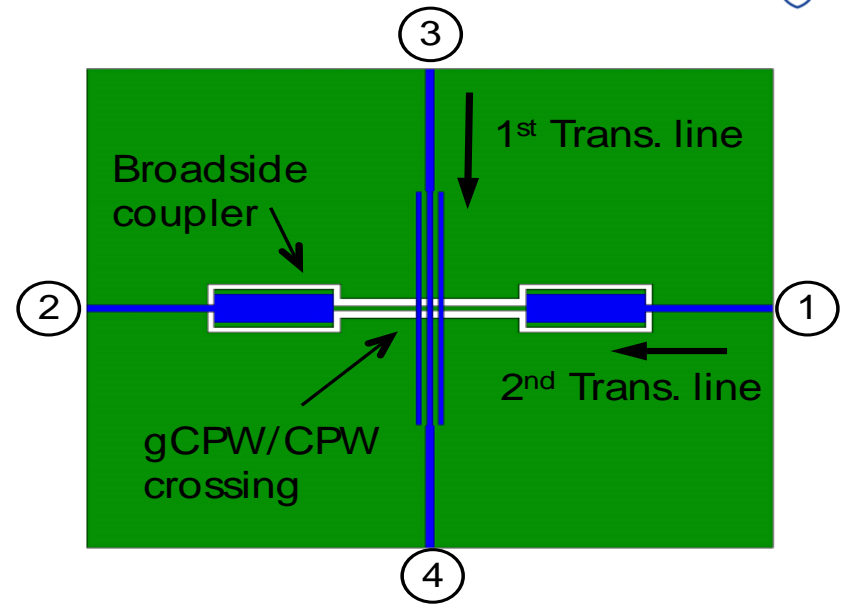
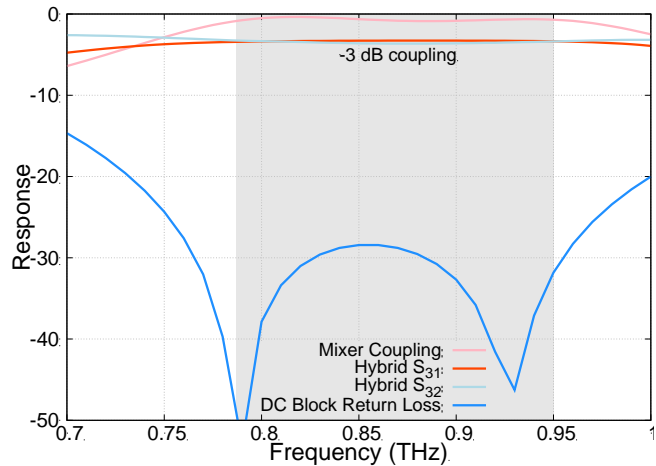
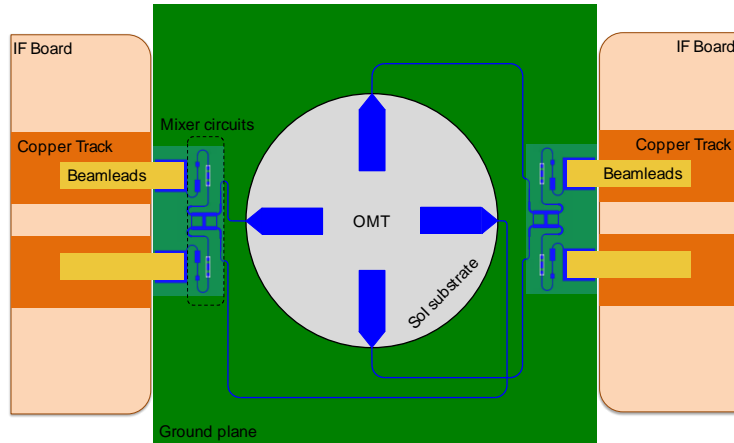
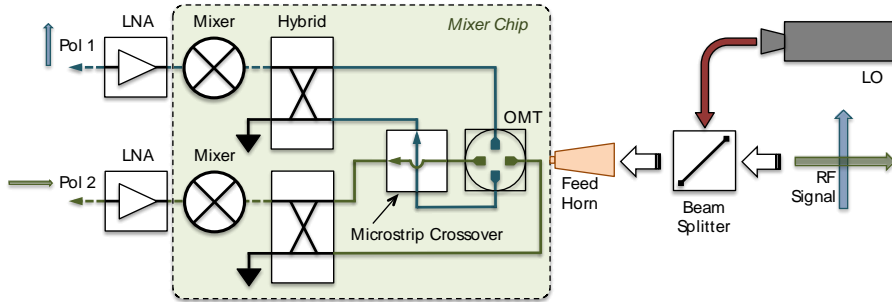
Credit: Dr. John Garrett (Oxford University)



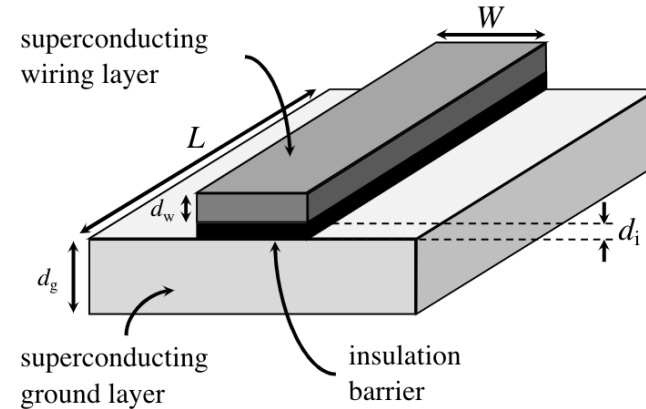
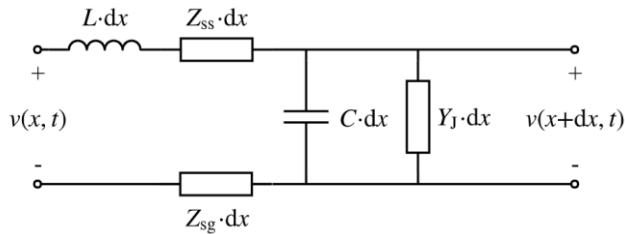
Planar beam splitter



On Chip dual polarization mixer



The Distributed SIS Junction



$$Z = W^{-1}(j\omega L_S + Z_{sw} + Z_{sg})$$

$$Y = W(j\omega C_S + Y_j)$$

$$Y_j = \frac{I_\omega}{V_\omega \cdot A}$$

$$\alpha = \frac{V_\omega}{V_{ph}}$$

$$\text{Re} \{I_\omega\} = \sum_{-\infty}^{\infty} J_n(\alpha) [J_{n-1}(\alpha) + J_{n+1}(\alpha)] \cdot I_{dc} (V_b + nV_{ph})$$

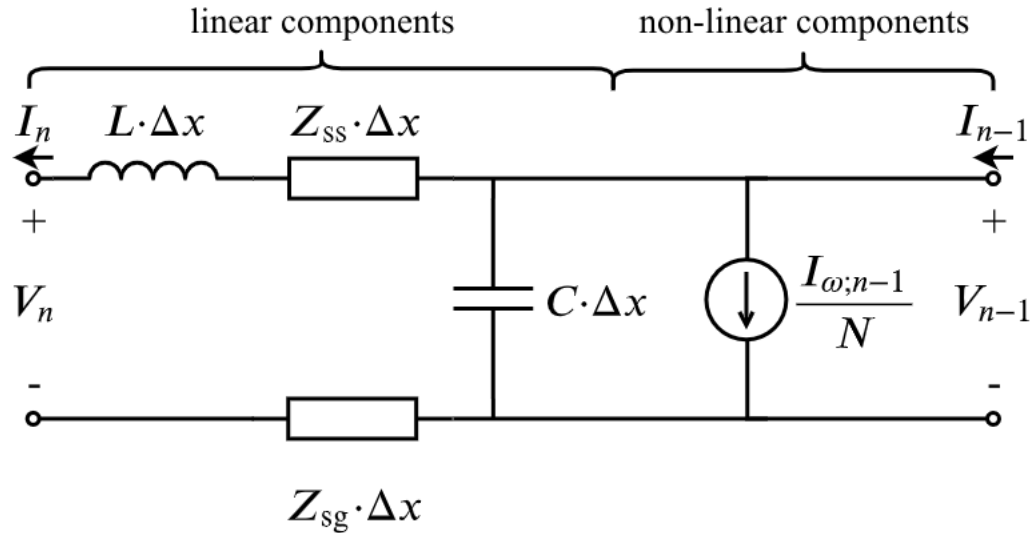
$$\text{Im} \{I_\omega\} = \sum_{-\infty}^{\infty} J_n(\alpha) [J_{n-1}(\alpha) - J_{n+1}(\alpha)] \cdot I_{KK} (V_b + nV_{ph})$$

$$\gamma = \sqrt{Z \cdot Y}$$

$$Z_0 = \sqrt{Z/Y}$$

$$Z_{in} = Z_0 \coth \gamma L$$

The nonlinear Model



$$\begin{bmatrix} V_n \\ I_n \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix}^{-1} \begin{bmatrix} V_{n-1} \\ I_{n-1} \end{bmatrix} - \begin{bmatrix} 0 \\ I_{\omega}/N \end{bmatrix}$$

$$\mathbf{V}^{k+1} = \mathbf{V}^k + [\mathbf{J}(\mathbf{V}^k)]^{-1} \cdot \Delta(\mathbf{V}^k)$$

Newton Raphson method

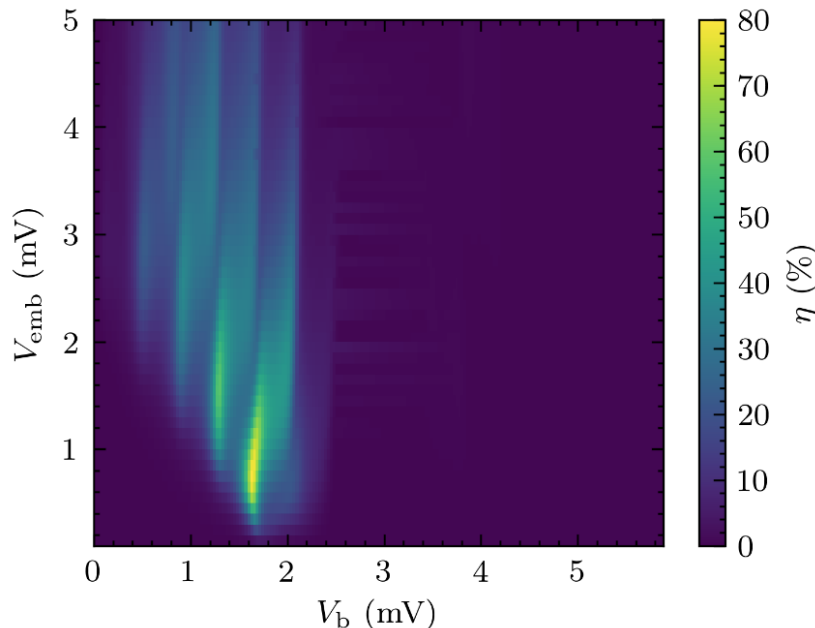
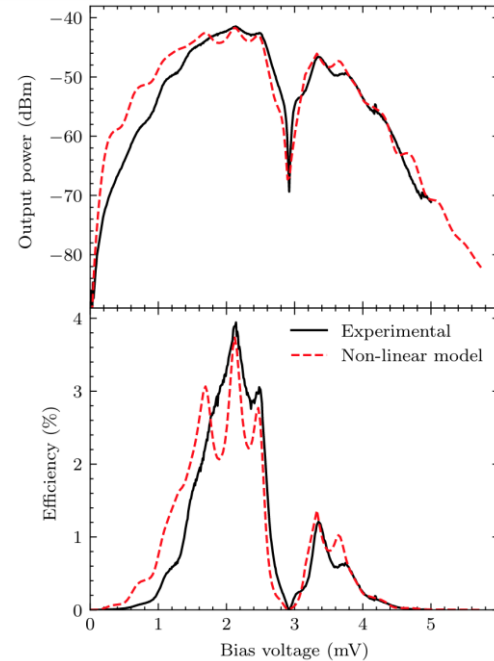
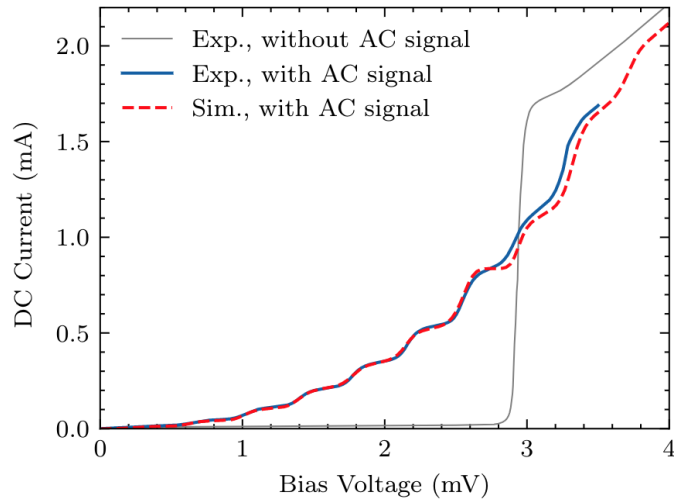
$$\Delta = INZ_{emb} - V_N - V_{emb}$$

$$\Delta' = I_N' \cdot Z_L' - V_N'$$

$$P_L' = \frac{|VN'|^2}{2\text{Re}\{Z_L'\}}$$

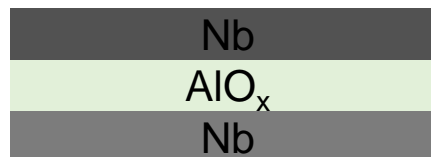
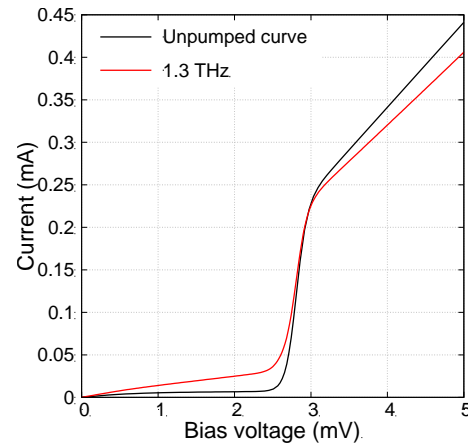
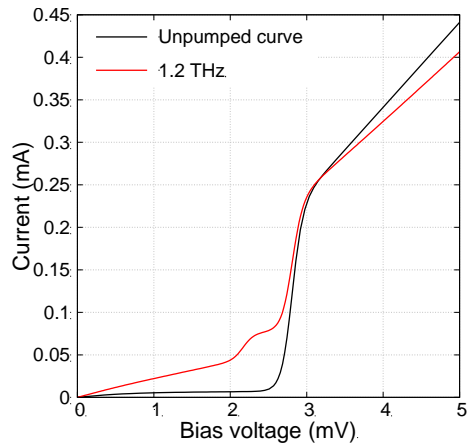
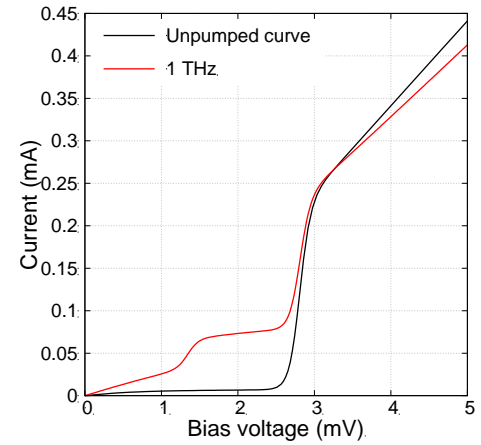
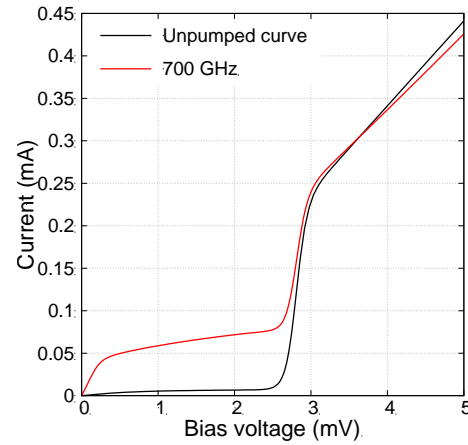
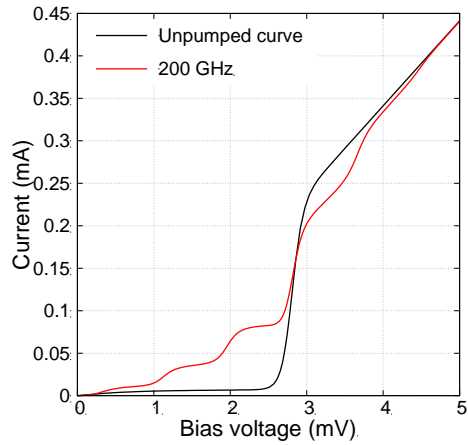
$$\eta = \frac{P_L'}{P_{emb}}$$

Results from the nonlinear model

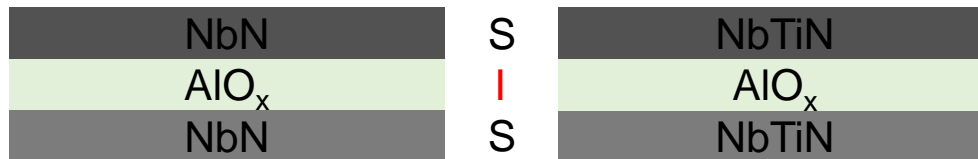
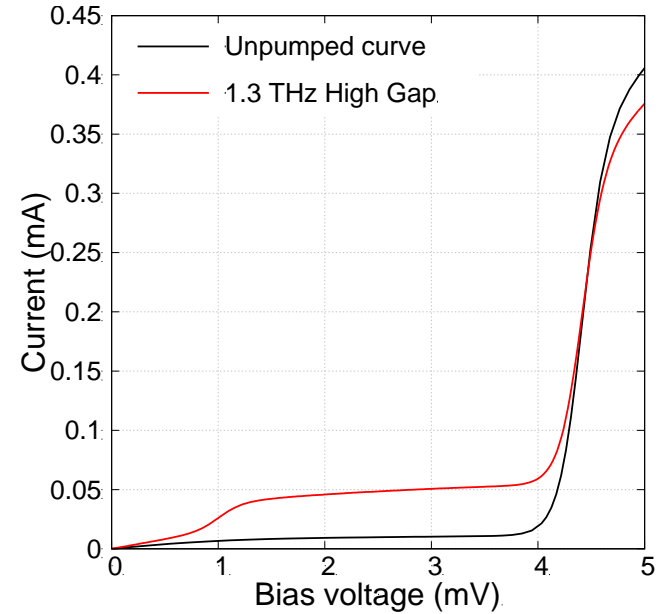
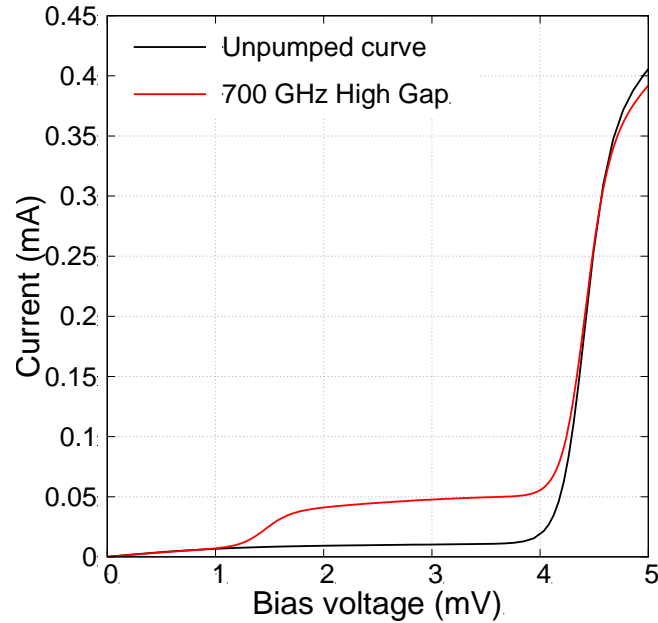


Efficiency of 80% at
 $(V_b, V_{emb}) = (1.62, 0.7)$ mV
Output power -38 dbm at
 $(1.24, 5.0)$ mV

THz Tunnel Junction

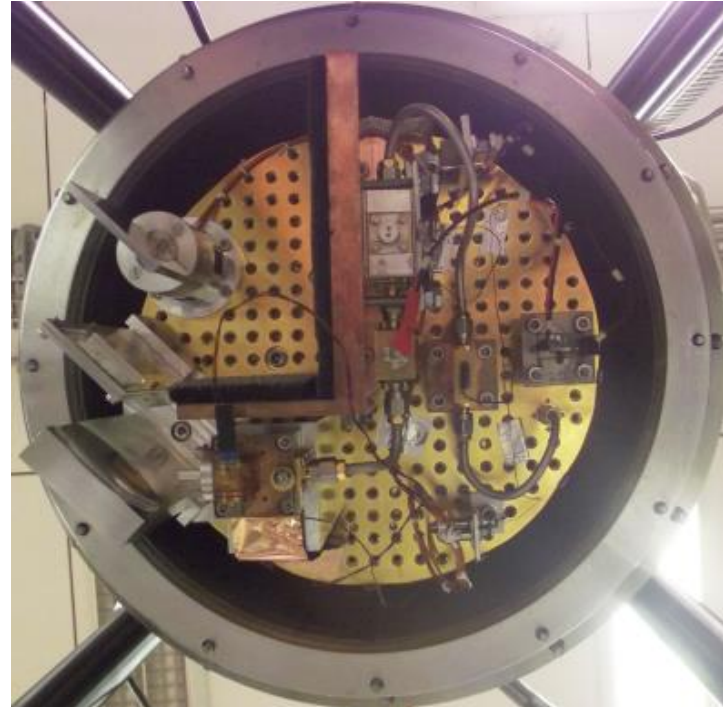
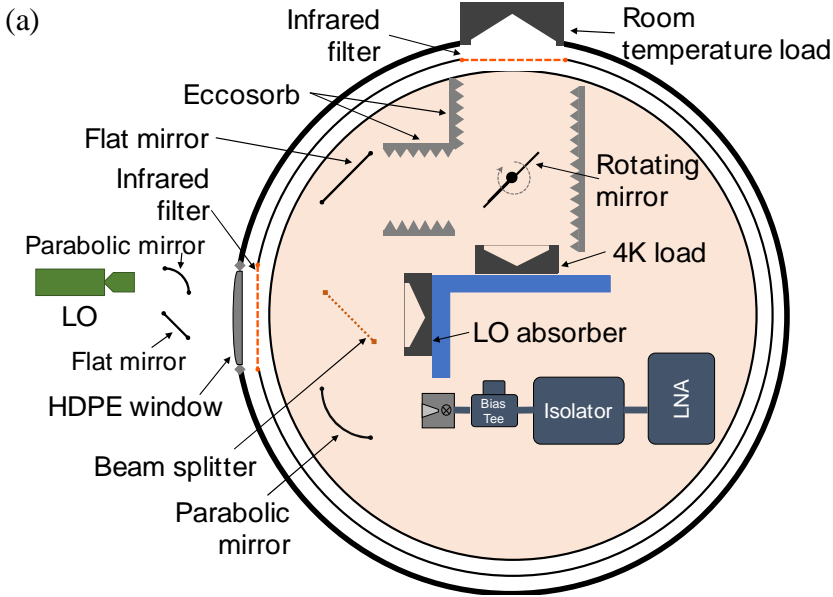


THz Tunnel Junction

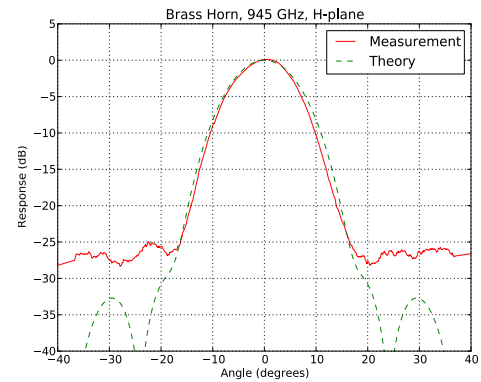
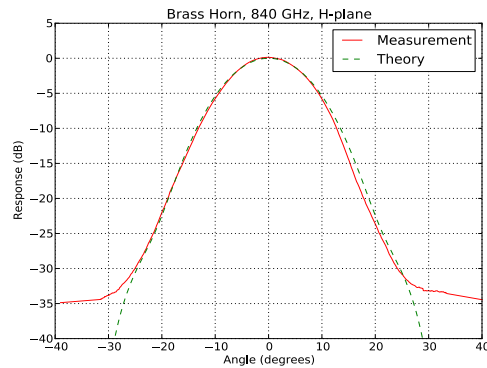
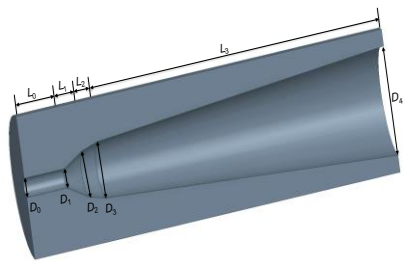


High gap superconductors

THz Test System



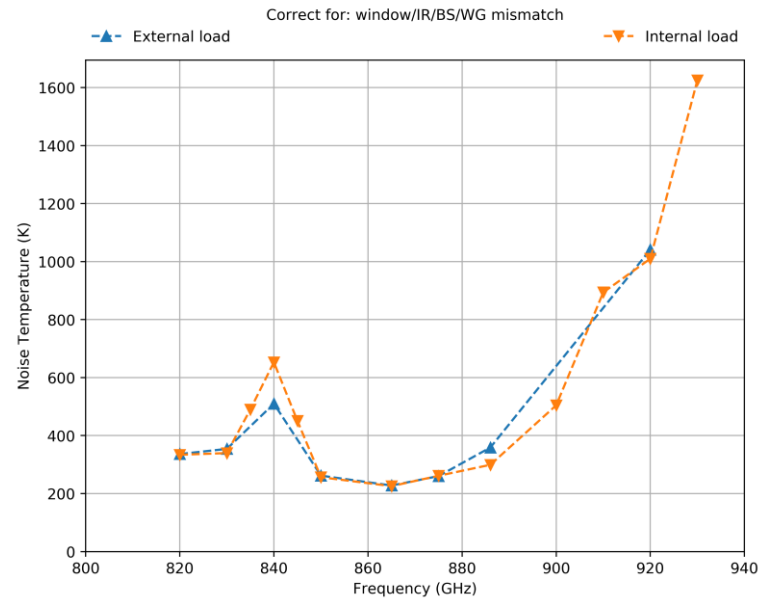
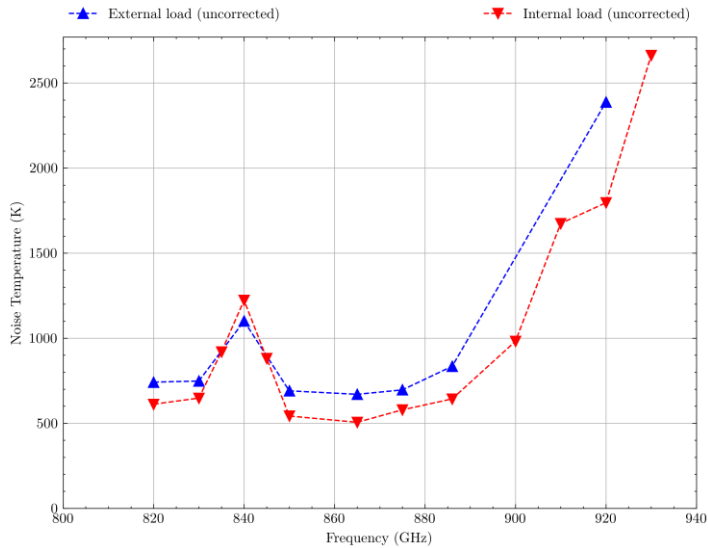
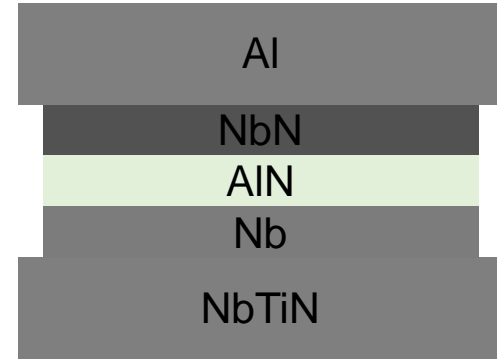
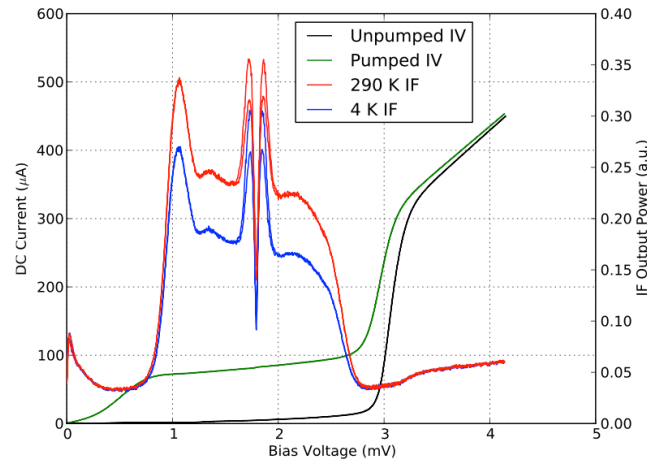
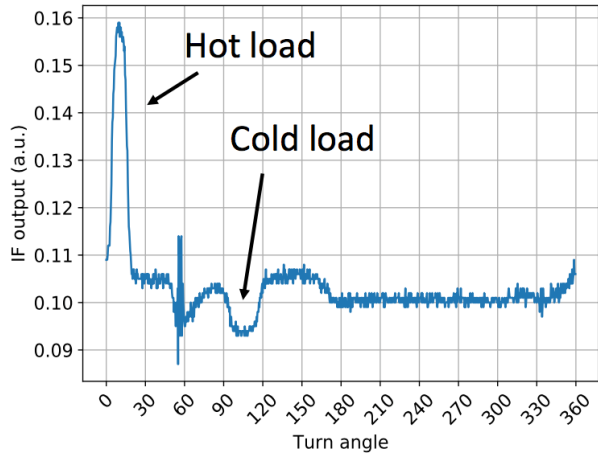
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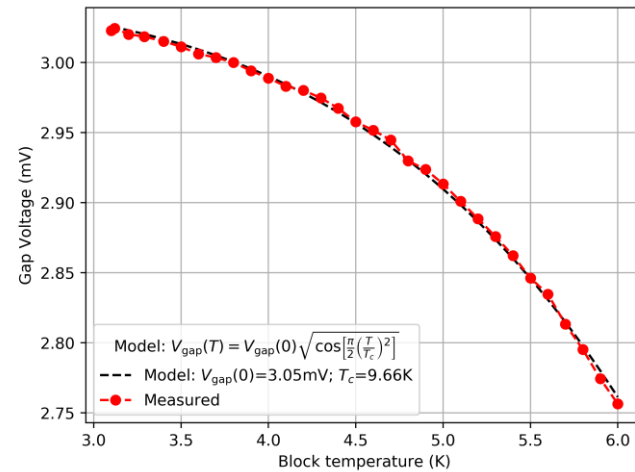
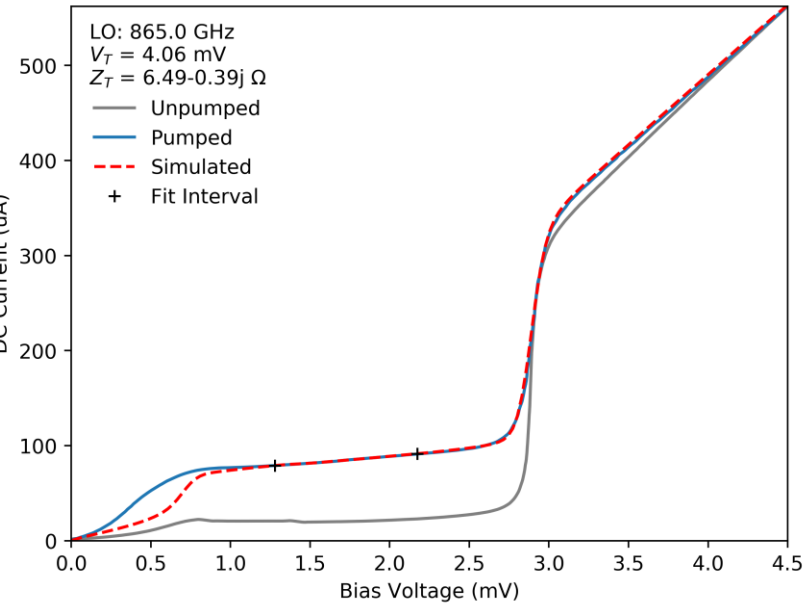
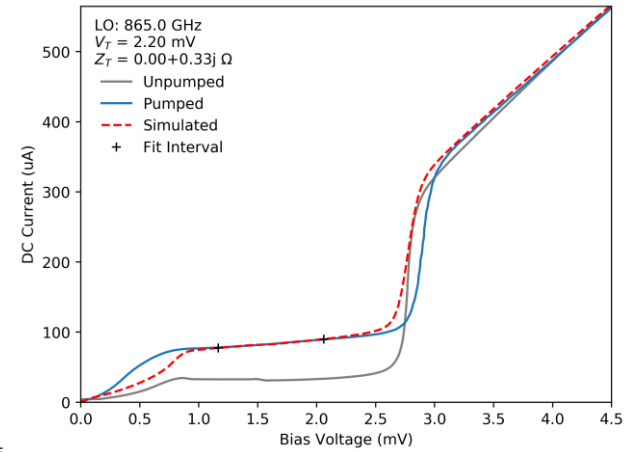
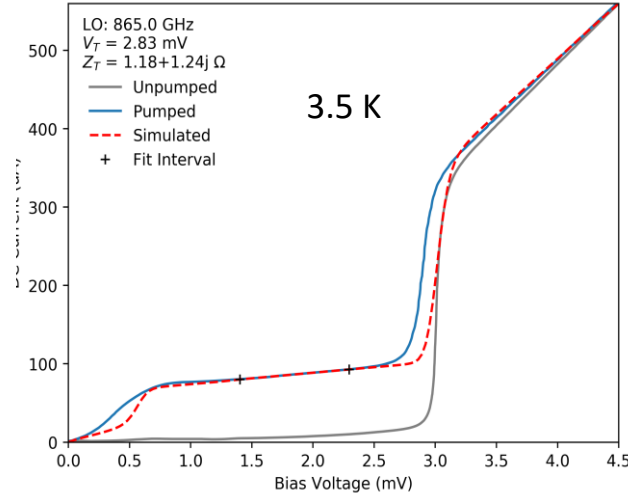
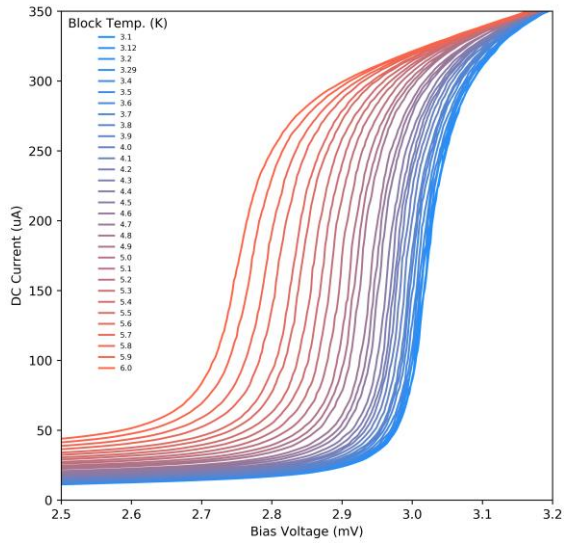
Supra-THz Mixer



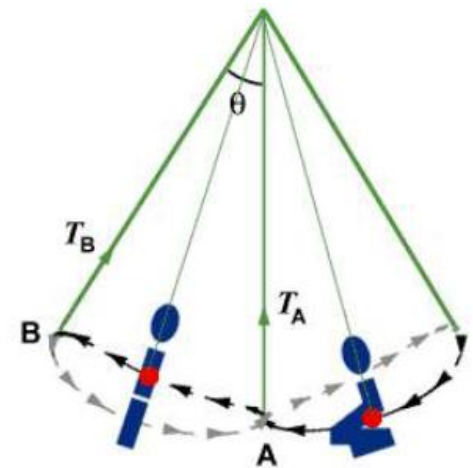
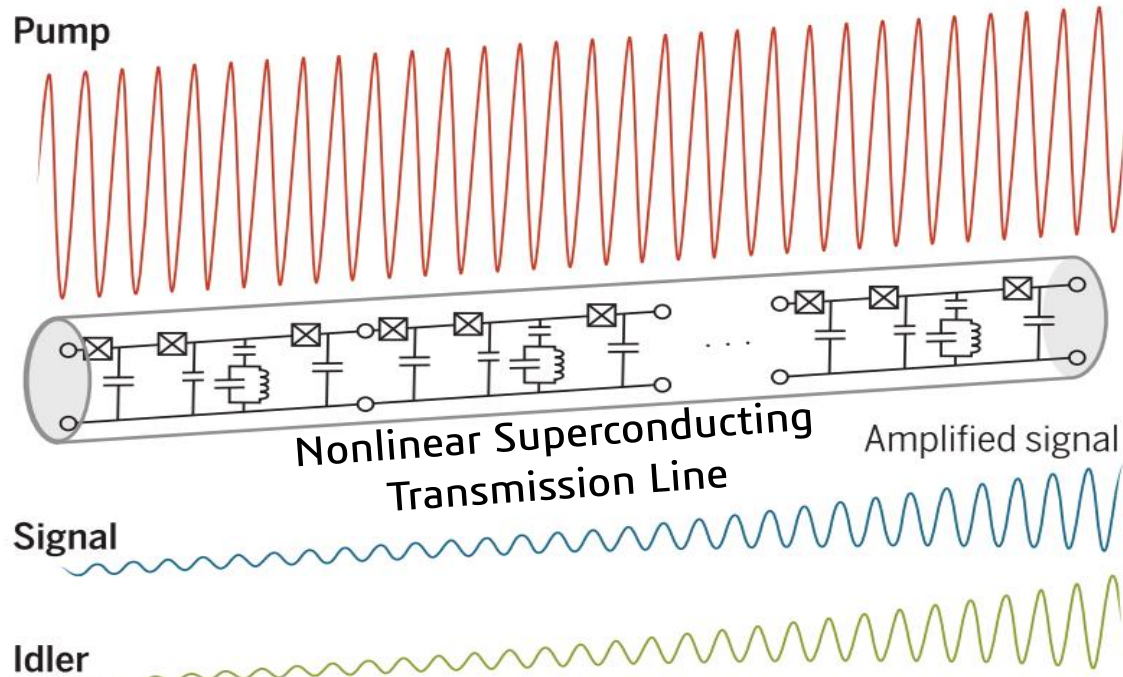
Collaboration project with University of Groningen



Supra-THz Mixer



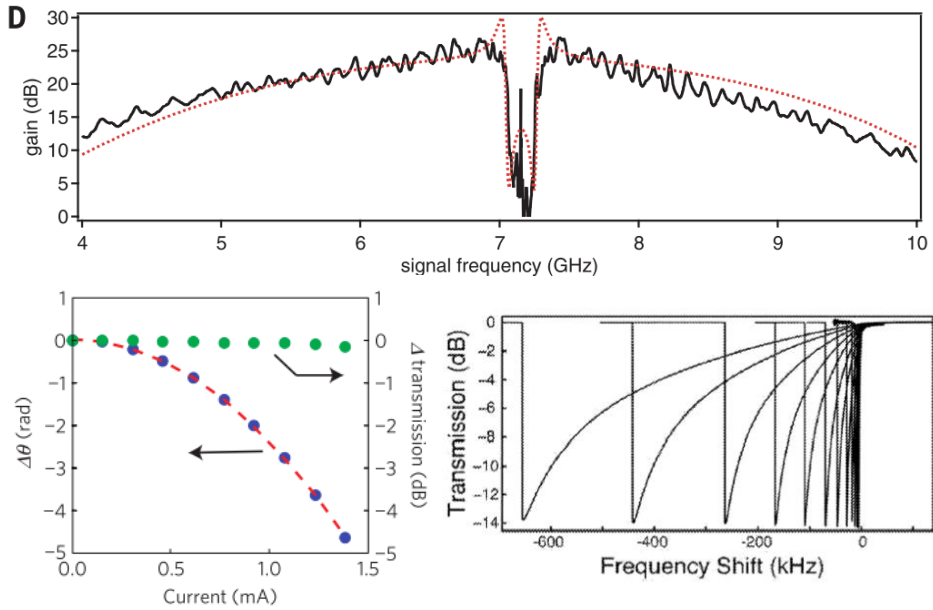
Superconducting Quantum Amplifiers



Berkeley, CA 94720,
Andrew N. Cleland Science 2015;350:280

Superconducting devices = quantum limited
Large interaction time between waves

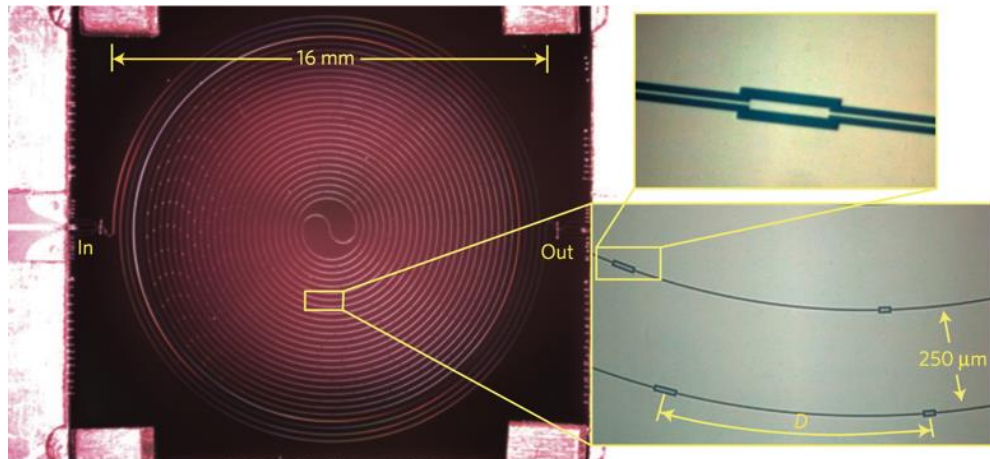
Travelling Wave Paramp – JJ & KI



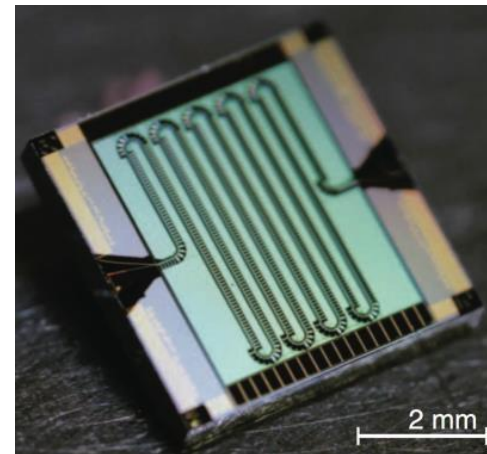
Bockstiegel et. al. J Low Temp Phys (2014) 176:476–482

Kinetic Inductance TWPA

Josephson Junctions TWPA



B. Eom Nature Physics 8, 623-627 (2012)



C. Macklin, Science, 350 (6258), p. 307, 2015

- High gain, broadband profile
- Low heat dissipation
- Quantum limited noise
- Compact
- Compatible with detector circuits

Phase Matching – Exp. Gain

Kevin O'Brien, Chris Macklin, Irfan Siddiqi, and Xiang Zhang. Resonant phase matching of josephson junction traveling wave parametric amplifiers. Physical review letters, 113(15):157001, 2014.
C. Macklin, Science, 350 (6258), p. 307, 2015

